

# **FINAL REPORT**

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## **PROGRAM TITLE:**

**VERY HIGH FREQUENCY MODULATION OF VERTICAL CAVITY  
LASERS WITH AN INTRACAVITY ABSORBER**

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## Executive Summary

Vertical cavity surface emitting lasers (VCSELs) have emerged as a key optical source in the digital optical communication domain as well as in RF photonics systems. Investigation to date has involved VCSELs with standard design of two distributed Bragg reflectors (DBR). Moreover, the specific design needs for high speed RF/linear Lightwave systems have been largely overlooked. In this program, we invented a new VCSEL structure with an integrated absorber that is uniquely well suited for many useful RF signal processing and transmission functions. This report describes results of experimental and theoretical investigation of the properties of such a VCSEL device. More details can also be found in the final report of the parent award, N00014-96-1-0583, "Very high frequency mode-stabilized VCSELs for linear/RF photonic applications."

The focus of this program is on designs and experiments of a three-contact vertical-cavity laser with an intracavity quantum-well. Because of the additional absorbing layer, this laser exhibits unusual device physics. Contrary to all existing laser-absorber integration, we achieved an independent control of the gain region and the quantum-well absorber. This allowed the device to be used as an integrated optical source and modulator, or alternatively, providing nonlinearity for self-pulsation and optical feedback.

An unprecedentedly wide range of applications is made possible with the unique properties and design flexibility offered by our three-contact, vertical-cavity laser with an intracavity absorber. We executed two distinct designs for our structures, leading to novel device physics and a wide range of applications, including low-chirp modulator-laser integration, QAM, bistable laser and self-pulsating laser.

Under this and the parent award programs, we graduated 3 PhD students, one post-doctoral researcher and one MS student. The grant also helped to start several new students. As the research scope became far wider than we initially anticipated, we are fortunate to be able to leverage the support from another ONR grant, MURI-RF Photonics program lead by UCLA. We published over 20 publications and presented about a dozen invited talks in international conferences.

This invention deserves more research work to fully exploit the opportunity brought forth by this new class of VCSELs we have started. In fact, other research organizations, e.g. Sandia National Labs and Tokyo Institute of Technology, have already followed our lead and published their new designs and inputs. We believe the impact of this work is an exceedingly far-reaching one.

# 1. VCSEL with Integrated Optical Modulator

## 1.1 Introduction

We have designed and fabricated a novel VCSEL with a quantum well (QW) absorber integrated into the upper mirror stack. The three-contact design of this structure enables independent control of the gain and absorber regions. Furthermore, our three-contact vertical cavity laser offers an additional degree of design flexibility that is not available in the analogous edge-emitters. Because the lasing wavelength of a VCSEL is determined simply by the optical cavity length, the position of the lasing wavelength relative to the absorber bandedge of our three-contact device can be precisely controlled. Exploiting this flexibility, we have executed two distinct designs for our structures, leading to novel device physics and a wide range of applications. A full report can be found in the report of the parent award, N00014-96-1-0583, "Very high frequency mode-stabilized VCSELs for linear/RF photonic applications." In this report we shall focus on our design for low-chirp optical intensity modulation and independent control of amplitude and phase modulation.

## 1.2 Baseline Design

Our novel device is a vertical cavity surface emitting laser with an n-p-n configuration that has an additional quantum well integrated into the upper mirror stack, as shown in Figure 1.

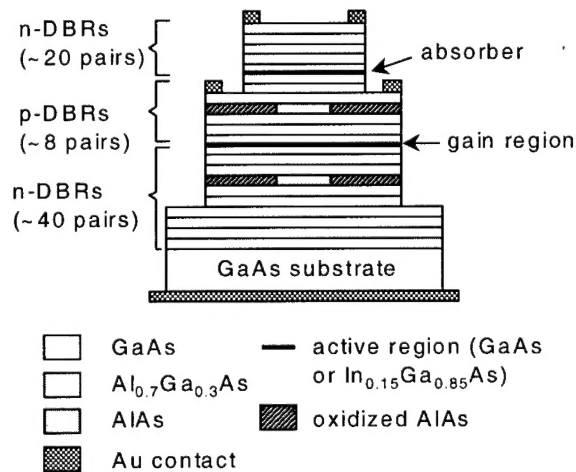


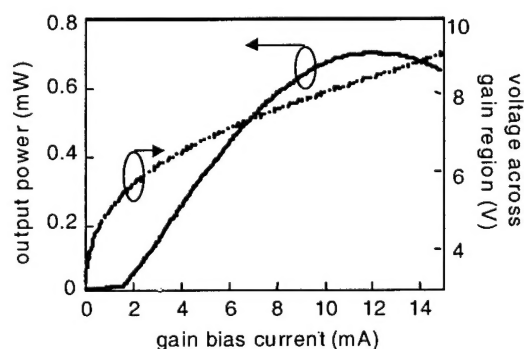
Figure 1. Device schematic.

A distributed Bragg reflector (DBR) mirror composed of n-doped  $\text{Al}_{0.7}\text{Ga}_{0.3}\text{As}$  and GaAs pairs is grown on a GaAs substrate. A single  $3\lambda/4$  AlAs oxidation layer is included in this DBR stack. This mirror is followed by the gain region, comprised of a  $1-\lambda$  thick  $\text{Al}_{0.35}\text{Ga}_{0.65}\text{As}$  spacer which contains either two 80 Å  $\text{In}_{0.15}\text{Ga}_{0.85}\text{As}$  quantum wells or three 70 Å GaAs quantum wells. A p-doped DBR stack with another  $3\lambda/4$  AlAs oxidation layer is grown on top of the gain region, followed by a  $5\lambda/4$  spacer with a single 80 Å  $\text{In}_{0.15}\text{Ga}_{0.85}\text{As}$  quantum well or

a 90 Å GaAs quantum well. An n-doped DBR mirror follows this second active region. During fabrication, the double mesa structure is obtained by dry etching, the two AIAs layers are exposed to wet thermal oxidation, and annular gold contacts are deposited as shown in Figure 1.

The n-p-n configuration of our VCSEL allows independent biasing of the two active regions. The lower multiple quantum well active region is forward biased to serve as the gain region, while the second, single-quantum well active region acts as an intracavity absorber under reverse bias. The two oxide layers improve the current and modal confinement of the device while inhibiting lasing in "oxide modes."

Typical CW lasing characteristics for our Design are shown in Figure 2. At 300 K with 0 V applied across the absorber section, the lasing threshold current of the device measured in Figure 2 is 1.7 mA, corresponding to a threshold current density of  $1.8 \text{ kA/cm}^2$ . At the lasing threshold, the voltage across the gain region is 5.6 V. The maximum output power of our devices ranges from 0.7-1 mW.



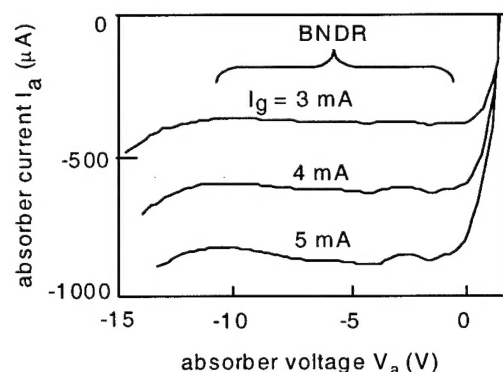
**Figure 2.** Typical CW lasing characteristics measured at 300 K, with 0 V across the absorber section.

The unique design flexibility of our VCSEL enables precise control of the relative positions of the lasing wavelength and absorber bandedge. Due to the characteristic short cavity length of a VCSEL, the lasing wavelength of our device is determined by the single Fabry-Perot mode which falls within the gain spectrum. Hence, the lasing wavelength can be precisely engineered by controlling the optical cavity length. To fully exploit this flexibility, we have fabricated devices with two distinct designs. In Design A, the lasing wavelength is designed to be shorter than the absorber bandedge. In Design B, the lasing wavelength is slightly longer than the absorber bandedge. As expected, the behavior of our device is strongly dependent on the position of the lasing wavelength relative to the absorber bandedge.

### 1.3 Design A: Red-tuned Absorber

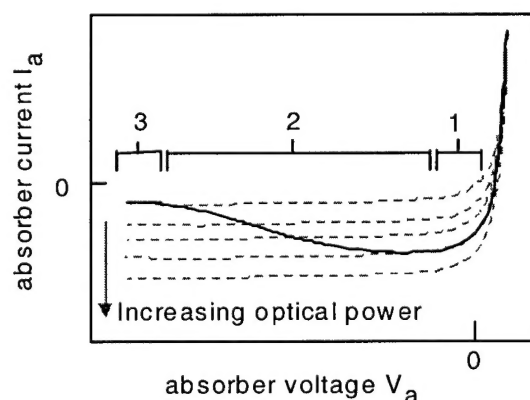
In Design A, the lasing wavelength is shorter than the absorber bandedge, that is the absorber is red-tuned. The lasing wavelength at twice the threshold injection current density is 829 nm, whereas the ground state excitonic peak of the absorber is positioned at 852 nm. The absorber in this device is a single 90 Å GaAs quantum well. The gain region is composed of three 70 Å GaAs quantum wells with the gain peak at 843 nm.

The current versus voltage (I-V) characteristic of the absorber in Design A is shown in Figure 3. The absorber in this Design Acts much like a standard photodetector, with the measured photocurrent increasing relatively linearly with increasing optical power in the cavity.



**Figure 3.** Experimental data of absorber current versus voltage ( $I_a$ - $V_a$ ) characteristics in a typical Design A, whose lasing wavelength is shorter than the absorber bandedge. The traces, each measured with a different DC gain bias current ( $I_g$ ), exhibit broad negative differential resistance (BNDR).

The broad region of negative differential resistance (BNDR) in the absorber characteristic arises due to variations in the cavity photon density in response to changes in the relative absorption; the origin of this phenomenon is shown schematically in Figure 4. The dashed curves in Figure 4 show the response of a typical photodiode; the solid line shows the response of our intracavity absorber. When the absorber is biased as shown in Region 1 (Figure 4), the absorber acts like a typical photodiode in the presence of the lasing photons. Applying a small reverse bias to the absorber (Region 2) causes increased absorption and hence a reduction in the cavity photon density, which has the net effect of reducing the absorber photocurrent. Under a strong reverse bias voltage (Region 3), the absorption is so large that the device ceases to lase, so the absorber behaves like a standard photodiode with very little incident optical power. Hence, it is this trade-off between the strength of the absorber and the cavity photon density which causes the BNDR in the absorber characteristic of Design A.



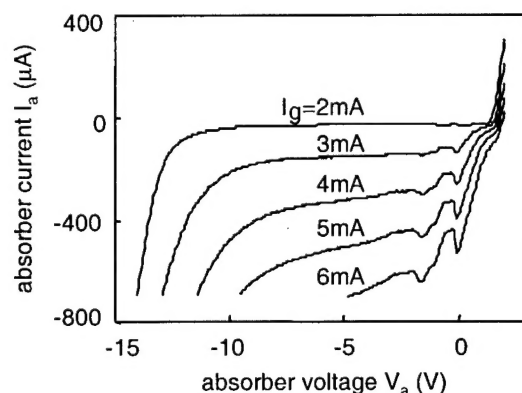
**Figure 4.** Schematic illustrating the origin of broad negative differential resistance in the absorber  $I_a$ - $V_a$  traces of Design A. Dashed lines show the response of a typical external photodiode; solid line shows the response of the intracavity absorber.

#### 1.4 Design B: Blue-tuned Absorber

In contrast, the lasing wavelength of Design B is slightly longer than the absorber bandedge. The lasing wavelength at twice the threshold injection current density is 959 nm, and the ground state excitonic peak of the absorber is positioned at 948 nm. The absorber is a

single 80 Å  $\text{In}_{0.15}\text{Ga}_{0.85}\text{As}$  quantum well. The gain region in this device is composed of two 80 Å  $\text{In}_{0.15}\text{Ga}_{0.85}\text{As}$  quantum wells with the gain peak at 940 nm.

The absorber I-V characteristic of Design B is shown in Figure 5. With this device design, a region of sharp negative differential resistance (SNDR) occurs in absorber I-V. We attribute this behavior to Stark-shifting of the ground-state absorption excitonic peak across the lasing wavelength.



**Figure 5.** Absorber current versus voltage ( $I_a$ - $V_a$ ) characteristics for Design B, in which the lasing wavelength is longer than the absorber bandedge. The traces, each measured with a different DC gain bias current ( $I_g$ ), exhibit sharp negative differential resistance (SNDR).

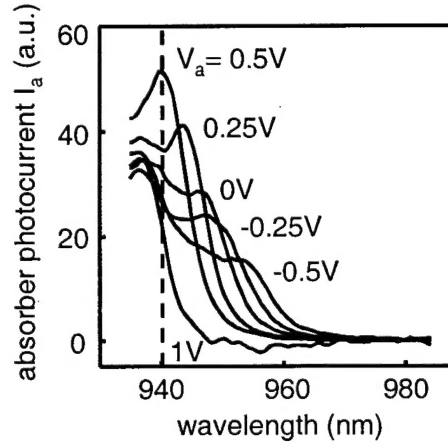
The measured absorption spectra for Design B are shown in Figure 6. Again, the ground state absorption excitonic peak Stark shifts with applied absorber bias voltage. We have modeled the expected shift in the absorption bandedge due to QCSE, using the techniques given in [1.9] with a 2400 Å long intrinsic region and a built-in voltage of 1.4 V.

If the wavelength of the injected light from the Ti-Sapphire probe laser is fixed at 940 nm (corresponding to the vertical line in Figure 6), then the absorber I-V characteristic, with no applied gain bias, exhibits SNDR as shown in Figure 7. The origin of this SNDR is illustrated by Figures 6 and 7. The absorber photocurrent is a maximum when the excitonic peak coincides with the 940 nm wavelength of the injected light; this occurs at an absorber bias voltage ( $V_a$ ) of 0.5V. As the reverse bias voltage is increased, the absorber photocurrent diminishes as the excitonic peak is Stark shifted away from 940 nm.

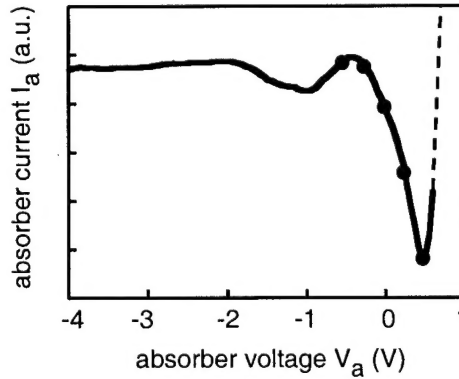
In order to extend this logic to explain the SNDR observed in the absorber when our device is lasing (Figure 5), the effects of the greatly increased cavity photon density must be considered. In the "cold cavity" measurements of Figures 6 and 7, in which our device is not lasing, the injected optical power is estimated to be on the order of 1 mW. However, when the device is lasing, the optical power in the cavity is on the order of 500 mW. The two-orders-of-magnitude increase in the cavity photon density when the device is lasing necessitates the inclusion of both carrier heating and carrier screening effects to explain its behavior.

Heating of the optical cavity at lasing powers causes red shifting of both the Fabry-Perot mode and of the gain and absorption spectra. At a 6 mA gain current bias, the CW lasing wavelength of Design B is 959.3 nm, which is 0.5 nm longer than the lasing wavelength under pulsed operating conditions (958.8 nm). Since the Fabry-Perot wavelength of a VCSEL has been shown to shift by 0.6 Å/°C and the absorption spectrum of bulk InGaAs shifts by 3 Å/°C, we estimate from the shift in lasing wavelength that with a 6 mA gain bias, the absorber bandedge redshifts by about 2 nm due to heating.





**Figure 6.** Solid lines show measured absorption spectra for Design B, with varying absorber bias voltages ( $V_a$ ) and zero gain bias current. Intersections between dashed line and absorption spectra give the relative absorption of light at 940 nm.



**Figure 7.** Absorber  $I_a$ - $V_a$  characteristic with 1 mW of 940 nm light injected from an external source. Gain bias current is zero. The dots correspond to the intersections in Figure 2.6 between the dashed line and the absorption spectra.

In addition, carrier screening causes band gap shrinkage of the absorber when the device is lasing. From the measured absorber photocurrent and a calculated carrier escape time of 10 ps, we estimate that at a gain bias current of 6 mA there are approximately  $10^{11}/\text{cm}^2$  carriers in the absorber quantum well. With this carrier density, we expect a band gap shrinkage of 10 meV, which causes a 7.5 nm red shift in the absorption spectrum.

If the absorption spectrum when the laser is operated at a gain bias current of 6 mA is red shifted by the expected 9.5 nm relative to the "cold cavity" case, then we expect the absorption excitonic peak with no applied absorber bias to occur at 957.5 nm. Since the lasing wavelength of Design B is about 959 nm, the absorption excitonic peak will be Stark shifted across the lasing wavelength with applied absorber bias voltage, thereby sweeping out the SNDR observed in Figure 5.

## 2. High Speed Digital Modulation

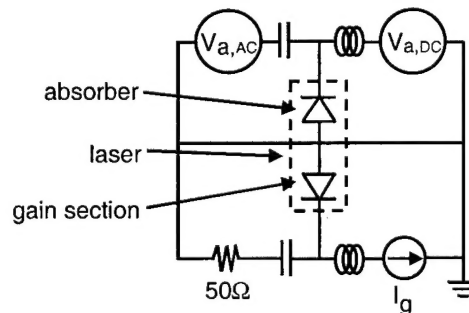
In this section, we present applications for Design A, which is ideal for use as an intracavity optical modulator. Optical intensity modulation using this technique is shown to provide

bandwidth and modulation efficiencies comparable to those obtained with direct gain modulation, while potentially reducing the frequency chirp.

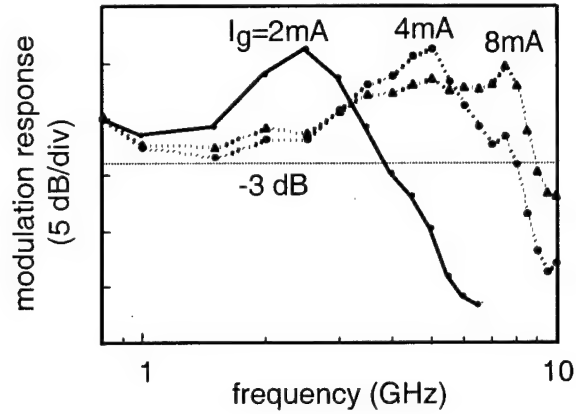
The ability to modulate vertical-cavity lasers at high frequencies is essential for data transmission applications. While direct gain modulation of VCSELs can be used to obtain reasonably large (GHz) modulation bandwidths, this technique is ultimately limited by frequency chirping and intrinsic parasitics [1, 2, 3]. Alternatively, an external modulator can be used to modulate the output power of a VCSEL without causing frequency chirping. However, this technique requires a long interaction length in order to achieve sufficient modulation depth, resulting in a cumbersome bulk. We demonstrate an alternative scheme in which the modulation signal is applied to the intracavity absorber of our novel VCSELs. By applying the modulation to our quantum-well absorber, which is located inside the optical cavity at a peak of the optical intensity distribution, we achieve high-speed modulation (9 GHz) and a high modulation efficiency (7 GHz/ $\sqrt{\text{mA}}$ ) with a single quantum well. Furthermore, small signal analysis of the laser rate equations predicts that at high modulation frequencies, absorber modulation causes less frequency chirp than does direct gain modulation [4, 5, 6].

The modulation speed of our absorber modulation technique was measured using the biasing circuitry shown in Figure 8. The gain region is biased with a simple DC current source, and a 50  $\Omega$  termination path is provided for any AC current arising in the gain region as the device is modulated. The absorber is reverse biased with a DC voltage source in parallel with a small-signal AC voltage modulation.

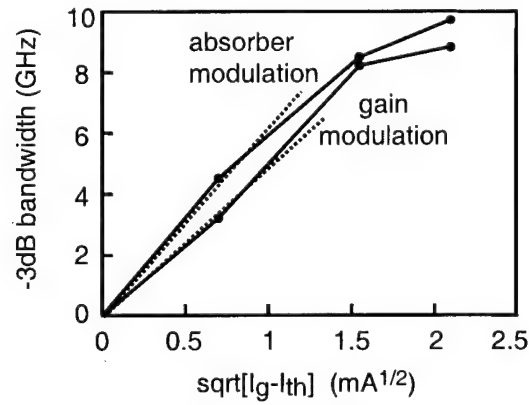
The response of the device to absorber modulation is shown in Figure 9, for a variety of gain bias conditions. We achieve a small signal modulation bandwidth of 9 GHz at a gain bias current of 8 mA and a modulation depth of 0.6%. As expected, the small-signal modulation bandwidth increases as the square root of the gain current bias above threshold. The bandwidth efficiency of the absorber modulation technique (7 GHz/ $\sqrt{\text{mA}}$ ) is shown in Figure 10 to be comparable to that obtained using direct gain modulation. Furthermore, the absorber modulation bandwidth is relatively independent of the choice of DC absorber bias; see Figure 11.



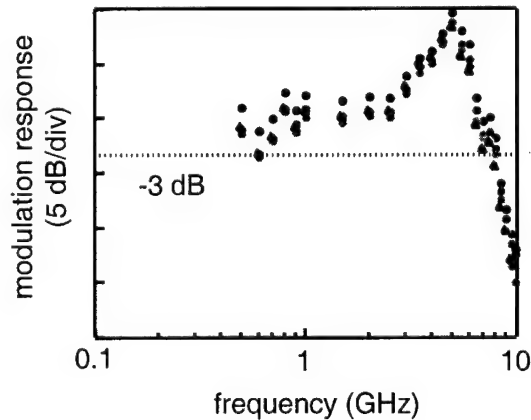
**Figure 8.** Biasing circuitry used for direct optical carrier modulation by small signal modulation of the intracavity absorber.



**Figure 9.** Absorber modulation response for varying DC gain bias currents ( $I_g$ ). The DC absorber bias  $V_{a,DC} = 1$  V, and the small-signal AC modulation of the absorber voltage is -10 dBm.



**Figure 10.** Bandwidth efficiency of absorber modulation compared to gain modulation. The slope efficiency (dashed line) is 7 GHz/ $\sqrt{\text{mA}}$  for absorber modulation and 5 GHz/ $\sqrt{\text{mA}}$  for gain modulation. Bias conditions for this measurement are  $V_{a,DC} = 1$  V and  $V_{a,AC} = -10$  dBm.



**Figure 11.** Absorber modulation response for fixed DC gain bias currents ( $I_g = 4$  mA) and varying DC absorber biases ( $V_{a,DC}$ ). The small-signal AC modulation of the absorber voltage is -10 dBm.

## 2.1 Frequency Chirp

The frequency chirp due to intracavity absorber modulation is compared in the following discussion to the chirp introduced by conventional direct current modulation [6].

Modulating the absorber effectively modulates the mirror loss since the absorber is part of the VCSEL mirror stack. This in turn leads to modulation of the photon lifetime, which causes variations in both the cavity photon density and the carrier density of the gain region. We use a small signal analysis of the rate equations (see also [7] for standard rate equations) to compare the frequency chirp due to intracavity absorber modulation ( $\delta\tau_p$ , where  $\tau_p$  is the photon lifetime) to the chirp obtained using direct current modulation ( $\delta J$ , where  $J$  is the injected current density). Define  $\delta n_g$  to be the change in the carrier density of the gain region when the absorber is modulated (i.e.  $\delta J = 0$ ). Likewise, define  $\delta n_J$  to be the change in the carrier density of the gain region when the pump current is modulated (i.e.  $\delta\tau_p = 0$ ). The ratio of the two,  $|\delta n_g / \delta n_J|$ , provides a comparison of the chirp produced by the two techniques. Note that this ratio depends on the modulation frequency  $\omega$ :

$$\frac{\delta n_g}{\delta n_J} = - \frac{\left( v_g g + S v_g \frac{\partial g}{\partial S} \right) \left( S v_g \Gamma \frac{\partial g}{\partial n} + 2 \Gamma \beta_{sp} B n \right)}{\left( i\omega + \frac{1}{\tau_{\Delta e}} + S v_g \frac{\partial g}{\partial S} \right) \left( i\omega - \Gamma v_g g + \frac{1}{\tau_p} - S v_g \Gamma \frac{\partial g}{\partial S} \right)} \quad (1)$$

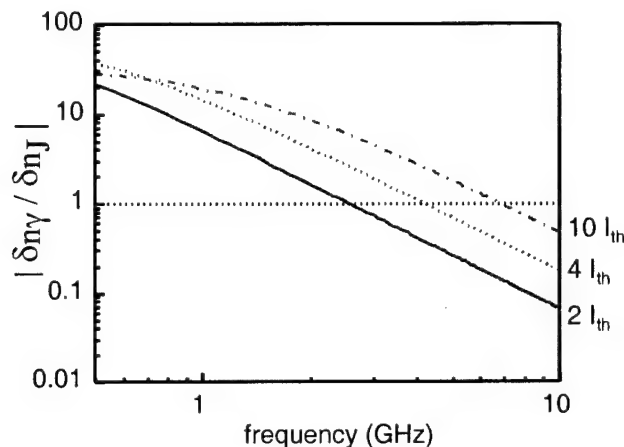
where  $v_g$  = group velocity,  $S$  = photon density,  $g$  = gain,  $\Gamma$  = confinement factor,  $\beta_{sp}$  = spontaneous emission factor,  $\tau_{\Delta e} = (A + 2Bn + 3Cn^2)^{-1}$ ,  $A$  = nonradiative recombination coefficient,  $B$  = radiative recombination coefficient, and  $C$  = Auger recombination coefficient.

The magnitude of the two-pole response given by (1) monotonically decreases with  $\omega$  due to the heavy damping introduced by  $\partial g / \partial S$  and  $\partial g / \partial n$ . For frequencies greater than  $f_c$  (the frequency at which  $|\delta n_g / \delta n_J| = 1$ ), absorber modulation produces less chirp than direct modulation. From equation (1):

$$\begin{aligned} \omega_c &= 2\pi f_c \approx \sqrt{S v_g \Gamma \frac{\partial g}{\partial n} \left( v_g g + S v_g \frac{\partial g}{\partial S} \right)} \\ &\approx \sqrt{\omega_R^2 + S^2 v_g^2 \Gamma \frac{\partial g}{\partial n} \cdot \frac{\partial g}{\partial S}} \end{aligned} \quad (2)$$

where  $\omega_R$  is the resonance frequency. Since  $\partial g / \partial S$  is negative,  $f_c$  is less than the resonance frequency. Figure 12 shows the calculated effect of absorber and gain modulation on the carrier density in the gain region for the same photon modulation depth and standard VCSEL parameters [7]. From Figure 9, the ratio of the -3dB-frequency  $f_{-3dB}$  to  $f_c$  is 1.6, so the critical frequency occurs well within the useful bandwidth of the device. Hence, absorber modulation at frequencies between  $f_c$  and  $f_{-3dB}$  results in less chirp than does direct gain modulation.

The carrier density within the absorber layer is also modulated. However, the absorber is far away from the active region, so any frequency chirping effects should be quite small.



**Figure 12.** Predicted effect of absorber and direct modulation on carrier density in the gain region. When  $|\delta n_\gamma| < |\delta n_J|$ , absorber modulation results in less chirp than does direct gain modulation.

## 2.2 Design Considerations

The relative placement of the gain peak wavelength ( $\lambda_{\text{gain}}$ ), the absorption peak wavelength ( $\lambda_{\text{abs}}$ ), and the Fabry-Perot transmission peak wavelength ( $\lambda_{\text{FP}}$ ) is an important design criterion for this device. Ideally, we should have  $\lambda_{\text{gain}} < \lambda_{\text{FP}}$  in order to achieve maximum overlap of the gain peak with the Fabry-Perot as the gain peak red-shifts with increasing pump current. Furthermore, we desire that  $\lambda_{\text{abs}} < \lambda_{\text{FP}}$  with these two wavelengths fairly close to each other, so that a small change in the applied voltage across the absorber induces a large change in the absorption due to the quantum-confined Stark effect. Since the wavelength of the pumped gain region will shift further than will the wavelength of the voltage-biased absorber, the ideal relative positioning of the three wavelengths should be  $\lambda_{\text{gain}} < \lambda_{\text{abs}} < \lambda_{\text{FP}}$ .

In contrast, if the ideal conditions are not met, we expect a reduction in modulation efficiency and speed. Placing the Fabry-Perot wavelength too much longer than the peak wavelength of the gain or absorber leads to poor modulation efficiency since greater voltages across the absorber are required to sweep the absorption edge across the Fabry-Perot wavelength [4]. Furthermore, the smaller overlap of the gain peak with the Fabry-Perot results in a narrow bandwidth [4].

In the device used for the experimental measurements, the Fabry-Perot wavelength (828 nm) is shorter than that of the gain (~843 nm) or absorber quantum well (~851 nm). In this case, changing the absorber DC bias does not result in a significant change in the absorption or the modulation response, as shown in Figure 11.

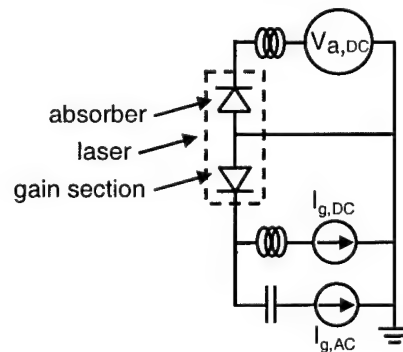
We expect that aligning the Fabry-Perot wavelength to be redder than the absorber quantum-well band edge will result in excellent modulation depth with this absorber modulation technique. A reverse bias across the absorber can then be used to sweep the absorption edge across the laser emission, resulting in a very high on-off ratio. Future work will include optimization of the bandwidth, growth and fabrication for large-signal modulation, and single-mode operation to demonstrate low frequency chirp.

### 3. Microwave Optical Subcarrier Modulation

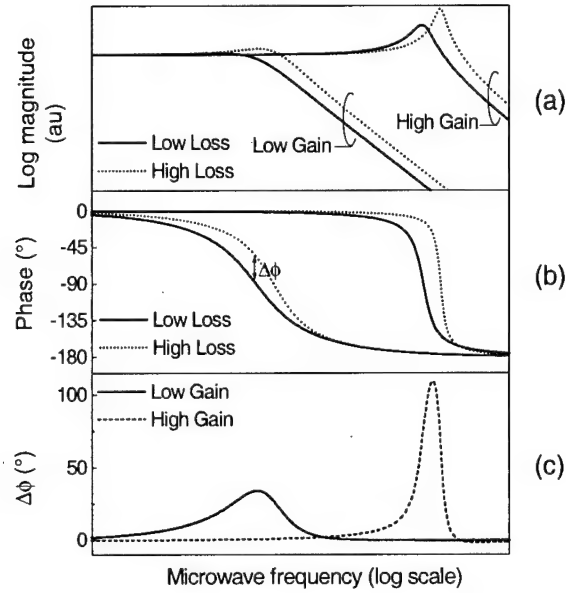
In systems with limited bandwidth, combinations of phase and amplitude modulation such as quadrature amplitude modulation (QAM) may be utilized to maximize the achievable bit rate. Use of our three-contact device for integrated generation and modulation of a microwave optical subcarrier enables independent control of both the phase and magnitude of the subcarrier. Generating phase and amplitude modulated signals for use in an optical system typically requires modulating an external microwave electrical signal with the required amplitude and phase, which is then used to modulate the optical source. In this program, we demonstrate a novel integrated arrangement, in which the phase and magnitude of a GHz-range optical intensity modulation of our three-contact VCSEL can be independently controlled by applying DC voltage and current biases to the VCSEL. In addition to applications in QAM optical data transmission, this device could also be used as an optical driver for phased antenna arrays [8, 9, 10].

A schematic of the biasing circuitry is shown in Figure 13. The three-contact, n-p-n configuration of our VCSEL allows independent control of the active absorber and gain regions. A GHz-range optical subcarrier is obtained by biasing the gain region above threshold with a DC current source ( $I_{g,DC}$ ) and applying a small-signal modulation ( $I_{g,DC}$ , typically  $\sim -10\text{dBm}$ ) with constant phase and amplitude to the gain region. The phase and amplitude of the resulting optical subcarrier can be independently controlled by varying the DC bias points of the gain current and absorber voltage.

The origin of the phase and amplitude variation of the subcarrier as the DC biases of the gain and absorber are varied can be understood qualitatively from the standard two-pole model of small-signal modulation of a semiconductor laser [7]. With this model, the relaxation oscillation frequency is given by  $\omega_R = (1/\tau_p v_g dg/dn P_0)^{1/2}$ , where  $\tau_p$  is the photon lifetime,  $v_g$  is the group velocity,  $dg/dn$  is the variation of the gain with the carrier density, and  $P_0$  is the steady-state photon density. Increasing the DC gain bias current increases the relaxation oscillation frequency and, for bias currents sufficiently close to threshold, reduces the damping of the response, as shown in Figures 14a and b. Varying the absorber bias voltage alters the magnitude of the absorption as discussed in Section 1 [11], hence changing both the photon lifetime and the steady state photon density. The resulting small changes in both the relaxation oscillation frequency and the damping of the small-signal modulation response cause shifts in both the magnitude and phase of the subcarrier, as shown schematically in Figure 14b. Figure 14c shows a schematic of the resulting phase shift between the high and low loss conditions used in Figures 14a and b, as a function of subcarrier frequency.



**Figure 13.** Biasing circuit used for phase and amplitude control of three-contact VCSEL.



**Figure 14.** Schematic of the 2-pole small-signal modulation frequency response. a) Subcarrier amplitude; b) Subcarrier phase; c) Expected phase shift  $\Delta\phi$  as absorber voltage is varied.

### 3.1 Experimental Results

Figure 15 shows the magnitude of the change in phase of the subcarrier as the absorber bias voltage is varied between 2V and -14V, for three different DC gain bias currents. These experimental results are in excellent agreement with the predictions of our model which are shown in Figure 14c. For each DC gain bias current, the phase shift is maximum when the subcarrier frequency coincides with the relaxation oscillation frequency, which increases as the square root of the current bias above threshold. Furthermore, as the gain bias current increases, the frequency range over which phase shifts occur narrows, as expected from Figure 14c.

With a standard two-contact laser such as that described by our simple two-pole model, the amplitude and phase responses of the subcarrier cannot be controlled independently of each other; see Figures 14a and b. However, our novel three-contact VCSEL, with separately biased gain and absorber regions, provides an extra degree of flexibility, allowing independent control of the subcarrier phase and magnitude. Figure 16 shows a contour plot for our device of the phase and magnitude of a 2.5 GHz optical subcarrier, as a function of the applied DC gain current and absorber bias voltage. The intersection of the magnitude contours with the phase contours demonstrates the ability to control the magnitude and phase independently. With this arrangement, we can obtain up to a 3.6dBm variation in the magnitude for a fixed phase of the subcarrier or we can vary the phase over 95° while maintaining a constant amplitude.

We have used this technique to demonstrate phase modulation of the optical subcarrier. With a DC gain bias ( $I_{g,DC}$ ) of 7.3 mA, we apply to the gain contact ( $V_a$ ) a 2 MHz, 5 V peak-to-peak square wave with a 5V DC offset. Under these conditions, the device operates at the two points shown in Figure 16; note that these two operating points have a constant amplitude but different phases. A time trace and the RF spectrum of the resulting phase-modulated optical subcarrier are shown in Figure 17. Analysis of the signal with a microwave

vector signal analyzer shows that the phase-modulated component of this signal is more than 18 dB greater than the amplitude-modulated component.

Furthermore, this device can be used for simultaneous phase and amplitude modulation. The signal space which can be achieved using our device is shown in Figure 3.11. Ideally, all of the first-quadrant signal points for 16-QAM should fall within this achievable signal space; the other three quadrants could then be obtained by applying standard  $90^\circ$  and  $180^\circ$  phase shifters to the microwave electrical drive current.

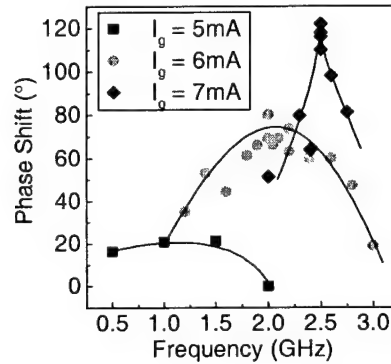


Figure 15. Measured subcarrier phase shift versus modulation frequency as absorber voltage is varied between 2V and -14V.

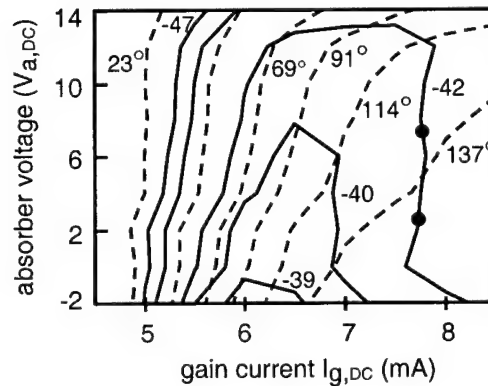
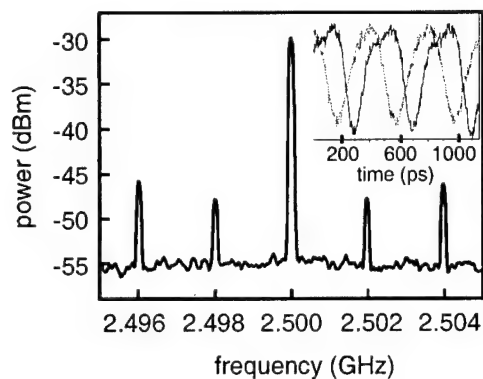
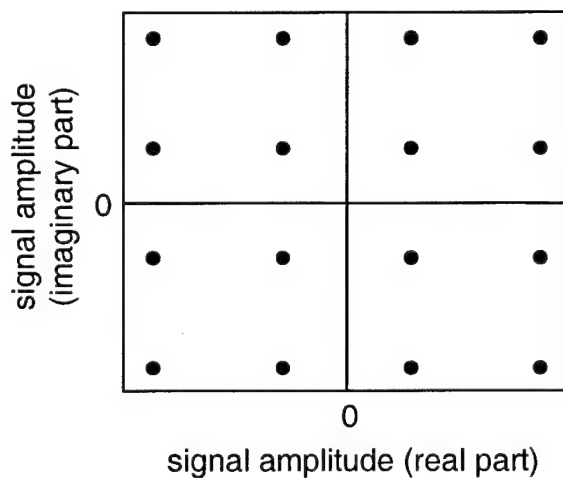


Figure 16. Contour plot of the measured phase and amplitude response of a 2.5 GHz optical subcarrier to tuning of the absorber bias voltage ( $V_{a,DC}$ ) and the gain bias current ( $I_{g,DC}$ ). Solid curves are surfaces of constant subcarrier amplitude (measured in dBm), and dashed curves are surfaces of constant subcarrier phase. Dots show operating points for phase modulation in Figure 3.10.





**Figure 17.** RF spectrum of phase modulation of the 2.5 GHz optical subcarrier. Inset shows phase-shifted time traces of the subcarrier. DC gain bias is 7.3mA; the electrical power of the applied modulation is -5dBm. A 5V peak-to-peak, 2 MHz square wave with a 5V DC offset is applied to the absorber.



**Figure 18.** Achievable signal space which can be addressed using this device (gray region). Plot is overlaid with the signal points for 16-QAM.

## 4. Impact

We have demonstrated started an entire class of VCSEL devices with an integrated modulator/absorber section to perform a variety of novel functions. The most intriguing part of this design is the possibility of separately engineer the lasing wavelength and absorption wavelength. With this independent control, which has never been achieved before with any other laser structures, we discovered a large number of device applications. We executed two distinct designs for our structures, leading to novel device physics and a wide range of applications, including low-chirp modulator-laser integration, QAM, bistable laser and self-pulsating laser.

The impact of this research can be seen in two folds. First, other research organizations, e.g. Sandia National Labs and Tokyo Institute of Technology, have followed our lead and published their new designs and inputs. Secondly, the two patents filed under this program has been licensed and will be developed into products. We believe the impact of this work is an exceedingly far-reaching one.

Under this program, we graduated 3 PhD students, one post-doctoral researcher and one MS student. The grant also helped to start several new students. As the research scope became far wider than we initially anticipated, we are fortunate to be able to leverage the support from another ONR grant, MURI-RF Photonics program lead by UCLA. We published over 20 publications and presented about a dozen invited talks in international conferences.

We believe a wide avenue of device research has just been opened by this work. To just name a few examples of the future work, we recommend further studies on integrated modulators to establish chirp-engineering capability. This capability shall be made possible for the first time using this device structure. More work will be essential for high frequency RF photonics applications. Though it is difficult to design a VCSEL with linear light-current characteristics, with an integrated absorber, one shall be able to engineer a desirable LI curve to meet application needs.

## 5. Future Work

This invention deserves more research work to fully exploit the opportunity brought forth by this new class of VCSELs we have started. First and most important, the design flexibility in the modulator section is shown to lead to low chirp modulator. We had not explored the possibility of designing strained QWs with a designed QW absorption spectrum to reduce the chirp even further or to generate a negative chirp. This can be an extremely exciting area for very high speed (>40Gbps) applications. Furthermore, a 4-contact structure to combine both design A and B type QWs can lead to far more interesting device combinations.

## 6. Students Supported By This Program

Former students:	Sui F. Lim, PhD 1998 Now at Agilent
	W. Yuen, PhD 1999 Now at Bandwidth9 Inc.
	Rob Stone, Postdoc, 1999 Now at Bandwidth9 Inc.
	Janice Hudgings, PhD 1999 Now at Mount Holyoke College
	Steve Chase, MS 2000 Now at Johns Hopkins University
Current Students:	Lukas Chrostowski, Chih-Hao Chang, Jacob Hernandez, P. C. Ku

## 7. Publications

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